A COMPUTATIONAL METHOD FOR A CLASS OF JUMP LINEAR QUADRATIC SYSTEMS

K. H. WONG, N. LOCK and K. KAJI¹

(Received 8 June 1993; revised 27 October 1993)

Abstract

A class of linear systems subject to sudden jumps in parameter values is considered. To solve this class of stochastic control problem, we try to seek the best feedback control law depending only on the measurable output. Based on this idea, we convert the original problem into an approximate constrained deterministic optimization problem, which can be easily solved by any existing nonlinear programming technique. An example is solved to illustrate the efficiency of the method.

1. Introduction

In [2, 3], the study of optimal control for a class of linear systems with random parameters was initiated. The class of stochastic optimal control problems considered in these references involved a jump linear quadratic (JLQ) system, where the parameters are Markov jump processes with finitely countable states, and where an optimal control law is sought with respect to the mathematical expection of a quadratic cost. In [8, 9, 10], it was discovered that the optimal control law for this type of problem consisted of a full state feedback regulator. In [4], the optimal control law for the linear quadratic (LQ) problem without jumps in the parameter was established. In [7], the linear quadratic Gaussian (LQG) problem, where the parameters were subjected to continuously acting disturbances modelled by Gaussian white noise, was considered. In [6], the optimal feedback control law depending only on measurable output rather than on the state of the system for the JLQ problem was established. This result was further improved when a JLQG regulator was designed in [5] where only part of the system state, the output, was measured.

This paper deals exactly with the same problem as that considered in [6]. By seeking the best feedback control law depending only on the measurable output, we

¹Dept of Computational and Applied Math., Univ. of the Witwatersrand, Johannesberg, South Africa. (© Australian Mathematical Society, 1995, Serial-fee code 0334-2700/95

convert the original problem into an equivalent standard constrained deterministic optimization problem, which can be easily solved by any nonlinear programming technique. This method is better than those considered in [4] and [6] which are based on the necessary condition of optimality, because their algorithms could not guarantee that the system would be stable in every mode during each iteration of the algorithm. As in [1], the gradients of both the performance index and the constraints are derived via the introduction of the Hamiltonian and co-state matrices. Finally, an example is solved to illustrate the efficiency of our method.

2. Statement of the problem

Consider the following jump linear system

$$\begin{cases} \dot{x}(t) = A(r(t))x(t) + B(r(t))u(t) & (1a) \\ y(t) = C(r(t))x(t), & (1b) \end{cases}$$

$$y(t) = C(r(t))x(t),$$
 (1b)

where $x(t) \in \mathbb{R}^n$ is the state vector, $u(t) \in \mathbb{R}^m$ is the control vector and $y(t) \in \mathbb{R}^r$ is the output vector. The entries of the matrices A(r(t)), B(r(t)) and C(r(t)) are random, because they depend on the system mode r(t). r(t) is a stochastic jump process with a finite valuation set $S = \{1, \ldots, N\}$. The dynamics of r(t) are given in terms of the transition probabilities

$$\operatorname{Prob}\{r(t+\Delta)=j \mid r(t)=i\} = \begin{cases} \pi_{ij}\Delta + O(\Delta) & \text{if } i \neq j, \\ 1+\pi_{ii}\Delta + O(\Delta) & \text{if } i=j. \end{cases}$$
(2)

The initial mode $r(t_0)$ is assumed to have a known distribution P_0 .

For the sake of simplicity, let $[A_i, B_i, C_i]$ denote the matrices [A(r(t)), B(r(t)),C(r(t))] when the system operates in the *i*th mode (that is, r(t) = i).

We now define a quadratic performance criterion as follows:

$$\bar{J}(u, t_0, x(t_0), r(t_0)) = E\left\{\frac{1}{2}\int_{t_0=0}^{\infty} \left[x(\tau)^{\mathsf{T}}Qx(\tau) + u(\tau)^{\mathsf{T}}Ru(\tau)\right]d\tau \left| (t_0, x(t_0), r(t_0)) \right\}$$
(3)

where E stands for the mathematical expectation and Q and R are, respectively, a semipositive and a positive symmetric weighting matrix.

From [8] and [10], it is clear that the optimal control u^* can be expressed in a closed-loop form

$$u^{*}(t) = U(x(t), r(t)), \qquad (4)$$

with gains involving a set of N coupled Riccati equations. However, in most problems, some of the required states or mode variables are not available to the designer. Thus, as mentioned in [6], the more practical control law is the linear output feedback mode dependent law, that is, the law

$$u^{*}(t) = -F(r(t)y(t))$$
(5)

which minimizes (3).

3. An approximate problem

First, we restrict the control to have the form of linear output feedback mode

$$u(t) = -F_i y(t) \quad \text{when } r(t) = i, \tag{6}$$

where $F_i(i = 1, ..., N) \in \mathbb{R}^{m \times r}$ is the gain matrix when the system is operating in mode *i*.

THEOREM 3.1. Assume that the system starts in mode i_0 and the initial state is x_0 . Then the performance criteria (3) can be reduced to

$$\bar{J} = \frac{1}{2} x_0^{\mathsf{T}} K_{i_0} x_0, \tag{7}$$

where $K_i \in \mathbb{R}^{n \times n}$ (when r(t) = i) satisfies the equation

$$K_{i}\tilde{A}_{i} + \tilde{A}_{i}^{\mathsf{T}}K_{i} + Q + C_{i}^{\mathsf{T}}F_{i}^{\mathsf{T}}RF_{i}C_{i} + \sum_{j=1}^{N}\pi_{ij}K_{j} = 0, \qquad i = 1, \dots, N$$
(8)

where

$$\tilde{A}_i = A_i - B_i F_i C_i, \qquad i = 1, \dots, N,$$
(9)

provided that the matrices

$$\tilde{A}_i = \left(\tilde{A}_i + \frac{1}{2}\pi_{ii}I\right), \qquad i = 1, \dots, N$$
(10)

are stable.

Although the result of Theorem 3.1 was taken from [6], no proof has been given in [6]. Thus, for the sake of convenience of the reader, we shall provide the proof of Theorem 3.1 in Appendix A.

Furthermore, by assuming that the initial state x_0 is a random vector uniformly distributed on the surface of the *n*-dimension unit sphere, the performance (7) can be reduced to

$$\bar{J} = 1/(2n) \operatorname{trace}(K_{i_0}).$$
 (11)

The performance degradation of \overline{J} with respect to the optimal performance under full state feedback J^* can be measured by the degree of suboptimality μ , where

$$\mu = (\bar{J} - J^*)/J^*.$$
(12)

From (12) and (11), it is clear that this suboptimality μ depends on the initial plant mode $r(t_0)$. By assuming that $r(t_0)$ has a uniform distribution, the average value for μ can be found. More precisely, if

$$P_{0i} = \operatorname{Prob}(r(t_0) = i) = 1/N,$$
 (13)

an average suboptimal criterion is defined by

$$\bar{J} = \frac{1}{2Nn} \sum_{i=1}^{N} \operatorname{trace}(K_i).$$
(14)

Furthermore, a sufficient condition for \bar{A}_i , i = 1, ..., N to be stable is that

$$\Lambda_i = \frac{1}{2} \left[-\bar{A}_i - \bar{A}_i^{\mathsf{T}} \right] \tag{15}$$

be positive definite. Thus, by Sylvester's criterion, it is necessary and sufficient to ensure that for each i = 1, ..., N, the determinants of all the principle minors of Λ_i are greater than or equal to $\varepsilon > 0$, where ε is some small positive number. Let the determinants of the *n* principle minors of Λ_i be denoted by $g_{i,1}, ..., g_{i,n}$ respectively. Then we have

$$g_{i,j} \ge \varepsilon, \qquad j = 1, \dots, n.$$
 (16)

Thus, we obtain a deterministic optimization problem (P) as follows:

$$(P) \begin{cases} \min \bar{J} = \frac{1}{2Nn} \sum_{i=1}^{N} \operatorname{trace} (K_i) \\ \text{subject to } g_{ij} \ge \varepsilon, \qquad i = 1, \dots, N; \ j = 1, \dots, n \end{cases}$$
(17)

where K_i are the solutions of (8). The problem (P) can be easily solved by any existing nonlinear programming software, such as NLQP.

In order to solve the problem (P), we need to calculate the gradient of \overline{J} and $g_{i,j}$ with respect to the gain matrix F_i . The method for calculating these gradients is given in Section 4.

4. Gradient of the cost functional and the constraints

To derive a formula for the gradient of the cost function \overline{J} and the constraints g_{ij} , we need to regard (8) as the state equation in the optimal control problem. We first introduce the Hamiltonian H and the co-state matrices $\lambda_k (k = 1, ..., N)$ for \overline{J} . Let

$$H = \frac{1}{2Nn} \sum_{i=1}^{N} \operatorname{trace} \left\{ K_i \right\}$$
$$+ \sum_{i=1}^{N} \operatorname{trace} \left\{ \lambda_i \left[K_i \tilde{A}_i + \tilde{A}_i^{\mathsf{T}} K_i + Q + C_i^{\mathsf{T}} F_i^{\mathsf{T}} R F_i C_i + \sum_{j=1}^{N} \pi_{ij} K_j \right] \right\}, (18)$$

where the co-state matrices λ_k $(k = 1, ..., N) \in \mathbb{R}^{n \times n}$ are symmetric, satisfying the following adjoint system:

$$\frac{\partial H}{\partial K_k} = 0 \qquad \text{implies} \qquad \frac{1}{2Nn}I + \lambda_k \tilde{A}_k^{\mathsf{T}} + \tilde{A}_k \lambda_k + \sum_{i=1}^N \pi_{ik}^T \lambda_i = 0.$$
(19)

THEOREM 4.1. The gradient of \overline{J} is given by

$$\frac{\partial \overline{J}}{\partial F_i} = \frac{\partial H}{\partial F_i}$$
$$= 2R F_i C_i \lambda_i C_i^{\mathsf{T}} \qquad (i = 1, \dots, N).$$
(20)

The proof of this theorem will be given in Appendix B.

REMARK 4.1. To calculate $\partial \overline{J}/\partial F_i$, we need to solve the system comprising of (8) and (19). This system is known as a system of Lyaponuv equations, which can be solved by existing software.

REMARK 4.2. The method for calculating the gradients of the constraints is exactly the same as that for calculating the gradients of \overline{J} .

5. Illustrative example

EXAMPLE 5.1. (Same as the example given in [6, page 899].)

Let $x \in R^2$, $u \in R$, $y \in R$. There are two modes, with the transition matrix $\Pi = \begin{pmatrix} 0 & 0 \\ 1 & -1 \end{pmatrix}$. In mode 1, the plant is governed by

$$A_{1} = \begin{pmatrix} -1 & 0.5 \\ 0 & 1 \end{pmatrix}, \qquad B_{1} = \begin{pmatrix} 0 \\ 1 \end{pmatrix}, \qquad C_{1} = (0, 1).$$
(21)

In mode 2, the plant is governed by

$$A_2 = \begin{pmatrix} 0.25 & 0.5 \\ 0 & 0.25 \end{pmatrix}, \qquad B_2 = \begin{pmatrix} 0 \\ 1 \end{pmatrix}, \qquad C_2 = (0, 1).$$

The criterion (3) is considered with unity weighting matrices Q and R.

From [6], we know that the optimal solution with full state feedback is given by:

$$K_{1} = \begin{pmatrix} 0.4949 & 0.1010 \\ 0.1010 & 2.4495 \end{pmatrix} \text{ and a gain matrix } F_{1} = (0.1010 & 2.4495);$$

$$K_{2} = \begin{pmatrix} 2.3548 & 0.5635 \\ 0.5635 & 1.768 \end{pmatrix} \text{ and a gain matrix } F_{2} = (0.5635 & 1.7688).$$

The technique employed in [6], which is based on the necessary condition of optimality, yields the following result:

$$K_{1} = \begin{pmatrix} 0.5 & 0.0995 \\ 0.0995 & 2.4503 \end{pmatrix} \text{ and a gain matrix } F_{1} = 2.5135;$$

$$K_{2} = \begin{pmatrix} 3.0 & 0.6577 \\ 0.6577 & 1.7967 \end{pmatrix} \text{ and a gain matrix } F_{2} = 1.9319.$$

The results obtained by solving the approximate problem (\overline{P}) is as follow:

$$K_{1} = \begin{pmatrix} 0.5 & 0.1010 \\ 0.1010 & 2.4497 \end{pmatrix} \text{ and a gain matrix } F_{1} = 2.4749;$$

$$K_{2} = \begin{pmatrix} 3.0 & 0.6583 \\ 0.6583 & 1.7967 \end{pmatrix} \text{ and a gain matrix } F_{2} = 1.9319.$$

Let J^* , J^0 and J^1 be the value of the objective function which corresponds to the full state feedback solution, the optimal output feedback obtained by Mariton and Bertrand [6] and the optimal output feedback obtained by our method respectively. Let μ^0 and μ^1 be the value of μ obtained by substituting $J = J^0$ and $J = J^1$ in (12) respectively. The performance can be summed up in Table 5.1, depending on the initial mode distribution

$$P_0 = \begin{pmatrix} P_{01} \\ 1 - P_{01} \end{pmatrix}$$

From the values of μ^0 and μ^1 in Table 5.1, it is clear that the expected performances obtained by our method is slightly better than those obtained by Mariton and Bertrand [6].

When $P_{01} = 1$, $r(t_0) = 1$, the result obtained by our method is 0.02% better than that obtained in [6]. When $P_{01} = 0.5$, Prob $(r(t_0) = 1) = 0.5$, the result obtained

<i>P</i> ₀₁	1	0.5	0
$8 \times J^*$	2.9444	3.6309	4.3168
$8 \times J^0$	2.9503	3.8735	4.7967
$8 \times J^1$	2.9497	3.8732	4.7967
$\mu^{0}(\%)$	0.20	6.68	11.12
$\mu^{1}(\%)$	0.18	6.67	11.12

TABLE 5.1. Comparison of optimal solution

by our method is 0.01% better than that obtained in [6]. When $P_{01} = 0$, $r(t_0) = 2$, the accuracy of the result obtained by our method and that obtained in [6] is exactly the same. However, for both our method and that used in [6], the worst performance $(P_{01} = 0, r(t_0) = 2)$ using output feedback (5) is only about 10% worse than the optimal expected performance using full state feedback.

Appendix A

PROOF OF THEOREM 3.1. From (3), (5), (1b), we get

$$\bar{J} = E\left\{\frac{1}{2}\int_{t_0=0}^{\infty} x(\tau)^{\mathsf{T}} \left[Q + C(r(\tau))^{\mathsf{T}} F(r(\tau))^{\mathsf{T}} R F(r(\tau)) C(r(\tau))\right] x(\tau) d\tau \\ \left| t_0, x(t_0), r(t_0) \right\}.$$
(A1)

In view of (8), (A1) can be simplified to

$$\bar{J} = E \left\{ \frac{1}{2} \int_{t_0=0}^{\infty} -x(\tau)^{\mathsf{T}} \left[\tilde{A}(r(\tau))^{\mathsf{T}} K(r(\tau)) + K(r(\tau)) \tilde{A}(r(\tau)) + \sum_{j=1}^{N} \pi_{r(\tau),j} K(r(\tau)) \right] x(\tau) d\tau \ \Big| \ t_0, x(t_0), r(t_0) \right\}.$$
(A2)

Now let

$$V(t) = x(t)^{\mathsf{T}} K(r(t)) x(t)$$
(A3)

and

$$V_i(t) = E\{V(t) \mid r(t) = i\}.$$
 (A4)

Then, from (1a), (5), (1b) and (9), we have

$$\dot{x}(t) = \tilde{A}(r(t))x(t). \tag{A5}$$

Thus, by differentiating (A4), we get

$$\dot{V}_i(t) = E\left\{x(t)^{\mathsf{T}} \left[\tilde{A}(r(t))^{\mathsf{T}} K(r(t)) + K(r(t))\tilde{A}(r(t)) + \dot{K}(r(t))\right] x(t) \mid r(t) = i\right\}.$$
(A6)

On the other hand,

$$E\left\{\dot{K}(r(t))|r(t)=i\right\} = \lim_{\Delta \to 0+} \frac{E\left\{K(r(t+\Delta))|r(t)=i\right\} - E\left\{K(r(t))|r(t)=i\right\}}{\Delta}.$$
(A7)

But

$$\lim_{\Delta \to 0+} E\left\{K(r(t+\Delta))|r(t) = i\right\} = K_i + \Delta \sum_{j=1}^N \pi_{ij}K_j + 0(\Delta) , \qquad (A8)$$

so that (A7) becomes

$$E\left\{\dot{K}(r(t))\Big|r(t)=i\right\} = \sum_{j=1}^{N} \pi_{ij}K_j.$$
 (A9)

Thus, from (A6), (A9), we get

$$\dot{V}_{i}(t) = E\left\{x(t)^{\mathsf{T}}\left[\tilde{A}(r(t))^{\mathsf{T}}K(r(t)) + K(r(t))\tilde{A}(r(t)) + \sum_{j=1}^{N}\pi_{ij}K_{j}\right]x(t) \middle| r(t) = i\right\}.$$
(A10)

From (A2) and (A10), we get

$$\bar{J} = E\left[\frac{1}{2}\int_{t_0=0}^{\infty} -\dot{V}(t)dt \mid t_0, x(t_0)r(t_0)\right]$$

= $\frac{1}{2}\left[x_0^{\mathsf{T}}K_{i_0}x_0 - E\left(V(\infty) \mid t_0, x(t_0), r(t_0)\right)\right].$ (A11)

Since $\tilde{A}_i + \frac{1}{2}\pi_{ii}I$ is stable for all *i*, we have

$$E\left\{V(\infty) \mid t_0, x(t_0), r(t_0)\right\} = 0.$$
 (A12)

The conclusion of this theorem follows easily from (A11) and (A12).

Appendix B

PROOF OF THEOREM 4.1. Let F_i be any parameter matrix when r(t) = i and let ΔF_i be any perturbation about F_i . Define

$$F_i(\varepsilon) = F_i + \varepsilon \Delta F_i. \tag{B1}$$

For brevity, let $K_k(\cdot)$ and $K_k(\cdot, \varepsilon)$ (k = 1, ..., N) denote, respectively, the solutions of (8) corresponding to parameter matrices $(F_1, \ldots, F_i, \ldots, F_N)$ and the perturbed parameter matrices $(F_1, \ldots, F_{i-1}, F_i + \varepsilon \Delta F_i, F_{i+1} \ldots, F_N)$ respectively. Let

$$\Delta K_k = \left. \frac{\mathrm{d} K_k(\cdot, \varepsilon)}{\mathrm{d} \varepsilon} \right|_{\varepsilon=0}, \quad (k = 1 \dots, N). \tag{B2}$$

Then, from (17), (19) and (8), we get

$$\Delta \overline{J} = \frac{d\overline{J}(F_i(\varepsilon))}{d\varepsilon} \bigg|_{\varepsilon=0}$$

= ΔH
= $\left\langle \frac{\partial H}{\partial F_i}, \Delta F_i \right\rangle + \sum_{k=1}^N \left\langle \frac{\partial H}{\partial K_k}, \Delta K_k \right\rangle,$ (B3)

where the inner product of two $r_1 \times r_2$ matrices

$$A = (a_{kl})$$
 and $B = (b_{kl})$

is defined by

$$\langle A, B \rangle = \sum_{l=1}^{r_2} \sum_{k=1}^{r_1} a_{kl} b_{kl}.$$
 (B4)

Hence from (19) and (B3), we get

$$\Delta \overline{J} = \left\langle \frac{\partial H}{\partial F_i}, \ \Delta F_i \right\rangle. \tag{B5}$$

Since ΔF_i is arbitrary, the conclusion of the theorem follows readily from (B5).

References

- [1] M. Athans, "The matrix maximum principle", J. Inform. Control 11 (1967) 592-606.
- [2] J. J. Florentin, "Optimal control of continuous time, Markov, stochastic systems", J. Electron. Control 10 (1961) 473–488.
- [3] N. N. Krasovskii and E. A. Lidskii, "Analytic design of controllers in systems with random attributes", J. Automat. Remote Control 22 (1961) 1021-1025, 1141-1146, 1289-1294.
- [4] W. S. Levine and M. Athans, "On the determination of the optimal constant output feedback gains for linear multivariable systems", *IEEE Trans. Automat. Control* AC-15 (1970) 44–48.
- [5] M. Mariton, "On the influence of noise on jump linear systems", IEEE Trans. Automat. Control AC-32 (1987) 1094–1097.

[10] A computational method for a class of jump linear quadratic systems

- [6] M. Mariton and P. Bertrand, "Output feedback for a class of linear system with stochastic jump parameters", *IEEE Trans. Automat. Control* AC-30 (1985) 898–900.
- [7] P. J. McLane, "Optimal stochastic control of linear systems with state and control dependent disturbances", *IEEE Trans. Automat. Control* AC-16 (1971) 793–798.
- [8] D. D. Sworder, "Feedback control of a class of linear systems with jump parameters", *IEEE Trans.* Automat. Control AC-14 (1969) 9–14.
- [9] W. M. Wonham, "On a matrix Riccati equation of stochastic control", SIAM J. Control Optim. 6 (1968) 681-697.
- [10] W. M. Wonham, "Random differential equations in control theory", in Prob. methods in applied maths. 2 (ed. A. T. Bharucha-Reid), (Academic, New York, 1970), 131-212.